

TPS22966 5.5-V, 6-A, 16-mΩ ON-Resistance Dual-Channel Load Switch

1 Features

- Input Voltage Range: 0.8 V to 5.5 V
- Integrated Dual-Channel Load Switch
- ON-Resistance
 - $R_{ON} = 16\text{ m}\Omega$ at $V_{IN} = 5\text{ V}$ ($V_{BIAS} = 5\text{ V}$)
 - $R_{ON} = 16\text{ m}\Omega$ at $V_{IN} = 3.6\text{ V}$ ($V_{BIAS} = 5\text{ V}$)
 - $R_{ON} = 16\text{ m}\Omega$ at $V_{IN} = 1.8\text{ V}$ ($V_{BIAS} = 5\text{ V}$)
- 6-A Maximum Continuous Switch Current per Channel
- Low Quiescent Current
 - 80 μA (Both Channels)
 - 60 μA (Single Channel)
- Low Control Input Threshold Enables Use of 1.2-V, 1.8-V, 2.5-V, and 3.3-V Logic
- Configurable Rise Time
- Quick Output Discharge (QOD)
- SON 14-Pin Package With Thermal Pad
- ESD Performance Tested per JESD 22
 - 2-kV HBM and 1-kV CDM

2 Applications

- Ultrabook™
- Notebooks and Netbooks
- Tablet PCs
- Consumer Electronics
- Set-top Boxes and Residential Gateways
- Telecom Systems
- Solid-State Drives (SSD)

3 Description

The TPS22966 is a small, low R_{ON} , dual-channel load switch with controlled turnon. The device contains two N-channel MOSFETs that can operate over an input voltage range of 0.8 V to 5.5 V and can support a maximum continuous current of 6 A per channel. Each switch is independently controlled by an on/off input (ON1 and ON2), which can interface directly with low-voltage control signals. In TPS22966, a 220- Ω on-chip load resistor is added for quick output discharge when switch is turned off.

The TPS22966 is available in a small, space-saving 2-mm \times 3-mm 14-SON package (DPU) with integrated thermal pad allowing for high power dissipation. The device is characterized for operation over the free-air temperature range of -40°C to 105°C .

Device Information⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)
TPS22966	WSON (14)	3.00 mm \times 2.00 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

Application Circuit

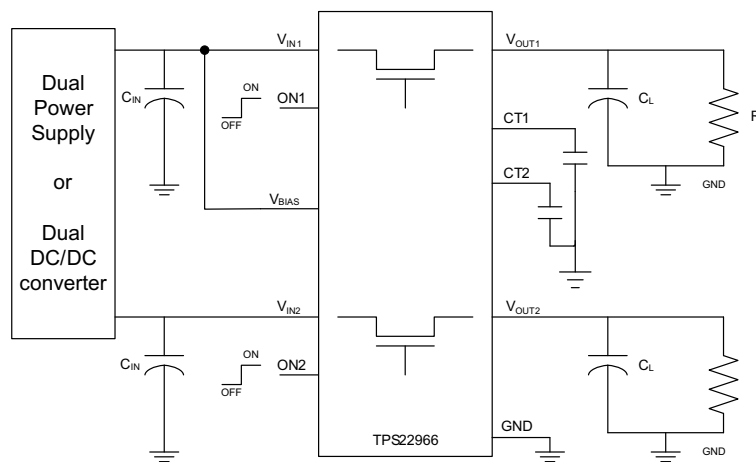


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4 Revision History

Changes from Revision D (January 2015) to Revision E Page

- Added temperature operating ranges to Electrical Characteristics (V_{BIAS} = 5.0 V) table. 5
- Added temperature operating ranges to Electrical Characteristics (V_{BIAS} = 2.5 V) table. 6
- Updated graphics in the Typical Characteristics section. 8

Changes from Revision C (June 2013) to Revision D Page

- Added *ESD Ratings* table, *Feature Description* section, *Device Functional Modes*, *Application and Implementation* section, *Power Supply Recommendations* section, *Layout* section, *Device and Documentation Support* section, and *Mechanical, Packaging, and Orderable Information* section 1

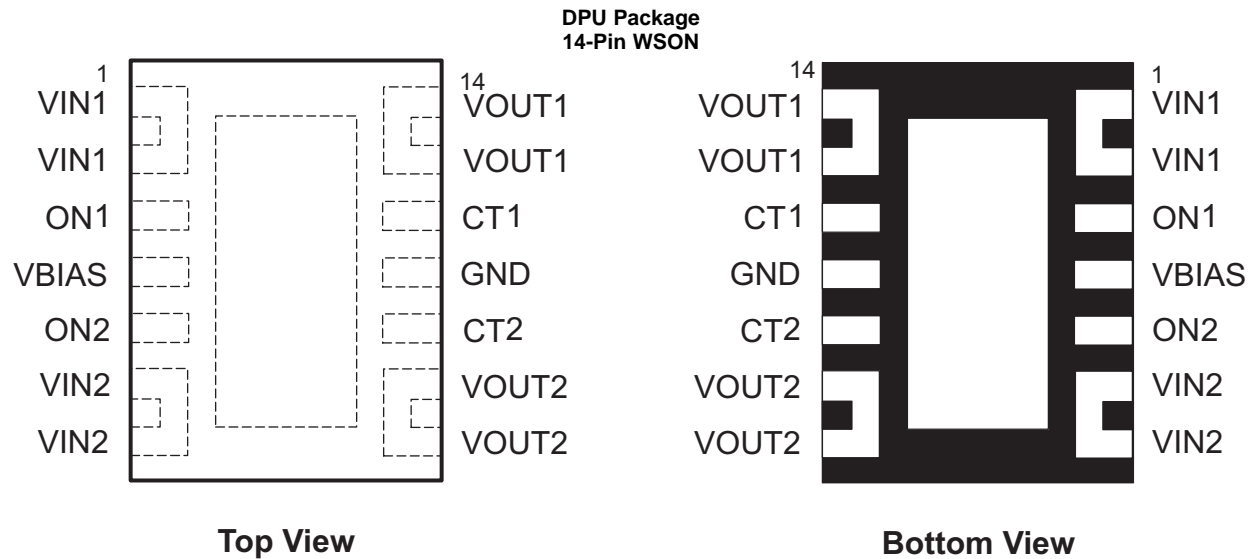
Changes from Revision B (December 2012) to Revision C Page

- Added VBIAS to ABSOLUTE MAXIMUM RATINGS table. 4
- Updated SWITCHING CHARACTERISTIC MEASUREMENT INFORMATION. 7
- Updated Test Circuit Diagram

Changes from Revision A (July 2012) to Revision B Page

- Updated Application Schematic. 1

5 Pin Configuration and Functions



Pin Functions

PIN		I/O	DESCRIPTION
NAME	NO.		
VIN1	1	I	Switch 1 input. Recommended voltage range for this pin for optimal R_{ON} performance is 0.8 V to V_{BIAS} . Place an optional decoupling capacitor between this pin and GND for reduce VIN dip during turn-on of the channel. See Application Information for more information.
VIN1	2	I	Switch 1 input. Recommended voltage range for this pin for optimal R_{ON} performance is 0.8 V to V_{BIAS} . Place an optional decoupling capacitor between this pin and GND for reduce VIN dip during turn-on of the channel. See Application Information for more information.
ON1	3	I	Active high switch 1 control input. Do not leave floating.
VBIAS	4	I	Bias voltage. Power supply to the device. Recommended voltage range for this pin is 2.5 V to 5.5 V. See Application Information .
ON2	5	I	Active high switch 2 control input. Do not leave floating.
VIN2	6	I	Switch 2 input. Recommended voltage range for this pin for optimal R_{ON} performance is 0.8V to V_{BIAS} . Place an optional decoupling capacitor between this pin and GND for reduce VIN dip during turn-on of the channel. See Application Information for more information.
VIN2	7	I	Switch 2 input. Recommended voltage range for this pin for optimal R_{ON} performance is 0.8 V to V_{BIAS} . Place an optional decoupling capacitor between this pin and GND for reduce VIN dip during turn-on of the channel. See Application Information for more information.
VOUT2	8	O	Switch 2 output.
VOUT2	9	O	Switch 2 output.
CT2	10	O	Switch 2 slew rate control. Can be left floating. Capacitor used on this pin should be rated for a minimum of 25 V for desired rise time performance.
GND	11	–	Ground
CT1	12	O	Switch 1 slew rate control. Can be left floating. Capacitor used on this pin should be rated for a minimum of 25 V for desired rise time performance.
VOUT1	13	O	Switch 1 output.
VOUT1	14	O	Switch 1 output.
Thermal Pad	–	–	Thermal pad (exposed center pad) to alleviate thermal stress. Tie to GND. See Layout for layout guidelines.

6 Specifications

6.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted)⁽¹⁾

		MIN	MAX	UNIT ⁽²⁾
V _{IN1,2}	Input voltage	-0.3	6	V
V _{OUT1,2}	Output voltage	-0.3	6	V
V _{ON1,2}	ON-pin voltage	-0.3	6	V
V _{BIAS}	V _{BIAS} voltage	-0.3	6	V
I _{MAX}	Maximum continuous switch current per channel		6	A
I _{PLS}	Maximum pulsed switch current per channel, pulse <300 μs, 2% duty cycle		8	A
T _J	Maximum junction temperature		125	°C
T _{LEAD}	Maximum lead temperature (10-s soldering time)		300	°C
T _{stg}	Storage temperature	-65	150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values are with respect to network ground terminal.

6.2 ESD Ratings

		VALUE	UNIT
V _(ESD)	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	±2000
		Charged-device model (CDM), per JEDEC specification JESD22-C101 ⁽²⁾	±1000

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

			MIN	MAX	UNIT
V _{IN1,2}	Input voltage range		0.8	V _{BIAS}	V
V _{BIAS}	Bias voltage range		2.5	5.5	V
V _{ON1,2}	ON voltage range		0	5.5	V
V _{OUT1,2}	Output voltage range			V _{IN}	V
V _{IH}	High-level input voltage, ON	V _{BIAS} = 2.5 V to 5.5 V	1.2	5.5	V
V _{IL}	Low-level input voltage, ON	V _{BIAS} = 2.5 V to 5.5 V	0	0.5	V
C _{IN1,2}	Input capacitor		1 ⁽¹⁾		μF
T _A	Operating free-air temperature ⁽²⁾		-40	105	°C

- (1) Refer to [Input Capacitor \(Optional\)](#).
- (2) In applications where high power dissipation and/or poor package thermal resistance is present, the maximum ambient temperature may have to be derated. Maximum ambient temperature [T_{A(max)}] is dependent on the maximum operating junction temperature [T_{J(max)}], the maximum power dissipation of the device in the application [P_{D(max)}], and the junction-to-ambient thermal resistance of the part/package in the application (θ_{JA}), as given by the following equation: T_{A(max)} = T_{J(max)} - (θ_{JA} × P_{D(max)})

6.4 Thermal Information

THERMAL METRIC		TPS22966		UNIT
		DPU		
		14 PINS		
$R_{\theta JA}$	Junction-to-ambient thermal resistance	52.3		°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	45.9		
$R_{\theta JB}$	Junction-to-board thermal resistance	11.5		
Ψ_{JT}	Junction-to-top characterization parameter	0.8		
Ψ_{JB}	Junction-to-board characterization parameter	11.4		
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	6.9		

6.5 Electrical Characteristics ($V_{BIAS} = 5.0\text{ V}$)

Unless otherwise noted, the specification in the following table applies where $V_{BIAS} = 5.0\text{ V}$. Typical values are for $T_A = 25^\circ\text{C}$. (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T_A	MIN	TYP	MAX	UNIT
POWER SUPPLIES AND CURRENTS							
$I_{IN(VBIAS-ON)}$	V_{BIAS} quiescent current (both channels)	$I_{OUT1} = I_{OUT2} = 0\text{ mA}$, $V_{IN1,2} = V_{ON1,2} = V_{BIAS} = 5.0\text{ V}$	-40°C to 105°C	80	120		μA
$I_{IN(VBIAS-ON)}$	V_{BIAS} quiescent current (single channel)	$I_{OUT1} = I_{OUT2} = 0\text{ mA}$, $V_{ON2} = 0\text{ V}$ $V_{IN1,2} = V_{ON1} = V_{BIAS} = 5.0\text{ V}$	-40°C to 105°C	60	120		μA
$I_{IN(VBIAS-OFF)}$	V_{BIAS} shutdown current	$V_{ON1,2} = 0\text{ V}$, $V_{OUT1,2} = 0\text{ V}$	-40°C to 105°C		2		μA
$I_{IN(VIN-OFF)}$	$V_{IN1,2}$ off-state supply current (per channel)	$V_{ON1,2} = 0\text{ V}$, $V_{OUT1,2} = 0\text{ V}$	$V_{IN1,2} = 5.0\text{ V}$	-40°C to 105°C	0.5	8	μA
			$V_{IN1,2} = 3.3\text{ V}$	-40°C to 105°C	0.1	3	
			$V_{IN1,2} = 1.8\text{ V}$	-40°C to 105°C	0.07	2	
			$V_{IN1,2} = 0.8\text{ V}$	-40°C to 105°C	0.04	1	
I_{ON}	ON pin input leakage current	$V_{ON} = 5.5\text{ V}$	-40°C to 105°C		1		μA
RESISTANCE CHARACTERISTICS							
R_{ON}	ON-state resistance (per channel)	$I_{OUT} = -200\text{ mA}$, $V_{BIAS} = 5.0\text{ V}$	$V_{IN} = 5.0\text{ V}$	25°C	16	19	$\text{m}\Omega$
				-40°C to 85°C		21	
				-40°C to 105°C		23	
			$V_{IN} = 3.3\text{ V}$	25°C	16	19	
				-40°C to 85°C		21	
				-40°C to 105°C		23	
			$V_{IN} = 1.8\text{ V}$	25°C	16	19	
				-40°C to 85°C		21	
				-40°C to 105°C		23	
			$V_{IN} = 1.5\text{ V}$	25°C	16	19	
				-40°C to 85°C		21	
				-40°C to 105°C		23	
			$V_{IN} = 1.2\text{ V}$	25°C	16	19	
				-40°C to 85°C		21	
-40°C to 105°C		23					
$V_{IN} = 0.8\text{ V}$	25°C	16	19				
	-40°C to 85°C		21				
	-40°C to 105°C		23				
R_{PD}	Output pulldown resistance	$V_{IN} = 5.0\text{ V}$, $V_{ON} = 0\text{ V}$, $I_{OUT} = 15\text{ mA}$	-40°C to 85°C	220	300	Ω	
			-40°C to 105°C		330		

6.6 Electrical Characteristics ($V_{BIAS} = 2.5\text{ V}$)

Unless otherwise noted, the specification in the following table applies where $V_{BIAS} = 2.5\text{ V}$. Typical values are for $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T_A	MIN	TYP	MAX	UNIT
POWER SUPPLIES AND CURRENTS							
$I_{IN(VBIAS-ON)}$	V_{BIAS} quiescent current (both channels)	$I_{OUT1} = I_{OUT2} = 0\text{ mA}$, $V_{IN1,2} = V_{ON1,2} = V_{BIAS} = 2.5\text{ V}$	-40°C to 85°C	32	37		μA
			-40°C to 105°C		40		
$I_{IN(VBIAS-ON)}$	V_{BIAS} quiescent current (single channel)	$I_{OUT1} = I_{OUT2} = 0\text{ mA}$, $V_{ON2} = 0\text{ V}$ $V_{IN1,2} = V_{ON1} = V_{BIAS} = 2.5\text{ V}$	-40°C to 105°C	23	40		μA
$I_{IN(VBIAS-OFF)}$	V_{BIAS} shutdown current	$V_{ON1,2} = 0\text{ V}$, $V_{OUT1,2} = 0\text{ V}$	-40°C to 105°C			2	μA
$I_{IN(VIN-OFF)}$	$V_{IN1,2}$ off-state supply current (per channel)	$V_{ON1,2} = 0\text{ V}$, $V_{OUT1,2} = 0\text{ V}$	$V_{IN1,2} = 2.5\text{ V}$	-40°C to 105°C	0.13	3	μA
			$V_{IN1,2} = 1.8\text{ V}$	-40°C to 105°C	0.07	2	
			$V_{IN1,2} = 1.2\text{ V}$	-40°C to 105°C	0.05	2	
			$V_{IN1,2} = 0.8\text{ V}$	-40°C to 105°C	0.04	1	
I_{ON}	ON pin input leakage current	$V_{ON} = 5.5\text{ V}$	-40°C to 105°C			1	μA
RESISTANCE CHARACTERISTICS							
R_{ON}	ON-state resistance	$I_{OUT} = -200\text{ mA}$, $V_{BIAS} = 2.5\text{ V}$	$V_{IN} = 2.5\text{ V}$	25°C	21	24	$\text{m}\Omega$
				-40°C to 85°C		27	
				-40°C to 105°C		29	
			$V_{IN} = 1.8\text{ V}$	25°C	19	22	
				-40°C to 85°C		25	
				-40°C to 105°C		27	
			$V_{IN} = 1.5\text{ V}$	25°C	18	21	
				-40°C to 85°C		24	
				-40°C to 105°C		26	
			$V_{IN} = 1.2\text{ V}$	25°C	18	21	
				-40°C to 85°C		24	
				-40°C to 105°C		26	
$V_{IN} = 0.8\text{ V}$	25°C	17	20				
	-40°C to 85°C		23				
	-40°C to 105°C		25				
R_{PD}	Output pulldown resistance	$V_{IN} = 2.5\text{ V}$, $V_{ON} = 0\text{ V}$, $I_{OUT} = 1\text{ mA}$	-40°C to 85°C	260	300		Ω
			-40°C to 105°C		330		

6.7 Switching Characteristics

PARAMETER		TEST CONDITION	MIN	TYP	MAX	UNIT
$V_{IN} = V_{ON} = V_{BIAS} = 5\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)						
t_{ON}	Turn-on time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		1310		μs
t_{OFF}	Turn-off time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		6		
t_R	V_{OUT} rise time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		1720		
t_F	V_{OUT} fall time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		2		
t_D	ON delay time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		460		
$V_{IN} = 0.8\text{ V}$, $V_{ON} = V_{BIAS} = 5\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)						
t_{ON}	Turn-on time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		550		μs
t_{OFF}	Turn-off time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		170		
t_R	V_{OUT} rise time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		325		
t_F	V_{OUT} fall time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		16		
t_D	ON delay time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		400		
$V_{IN} = 2.5\text{ V}$, $V_{ON} = 5\text{ V}$, $V_{BIAS} = 2.5\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)						
t_{ON}	Turn-on time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		2050		μs
t_{OFF}	Turn-off time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		5		
t_R	V_{OUT} rise time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		2275		
t_F	V_{OUT} fall time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		2.5		
t_D	ON delay time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		990		
$V_{IN} = 0.8\text{ V}$, $V_{ON} = 5\text{ V}$, $V_{BIAS} = 2.5\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)						
t_{ON}	Turn-on time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		1300		μs
t_{OFF}	Turn-off time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		130		
t_R	V_{OUT} rise time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		875		
t_F	V_{OUT} fall time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		16		
t_D	ON delay time	$R_L = 10\ \Omega$, $C_L = 0.1\ \mu\text{F}$, $C_T = 1000\ \text{pF}$		870		

6.8 Typical Characteristics

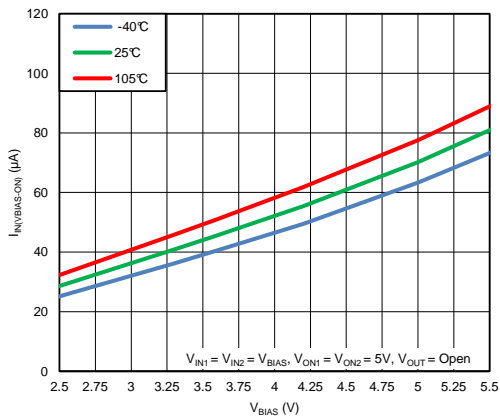


Figure 1. Quiescent Current vs. V_{BIAS} (Both Channels)

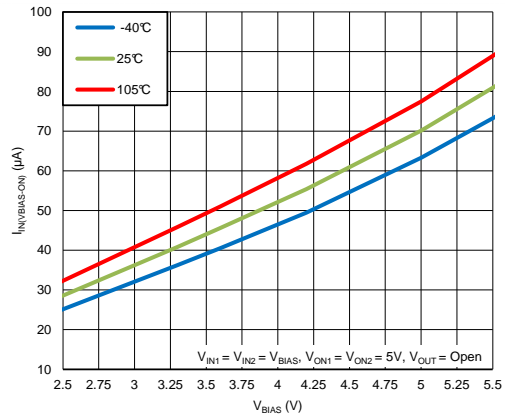


Figure 2. Quiescent Current vs. V_{BIAS} (Single Channel)

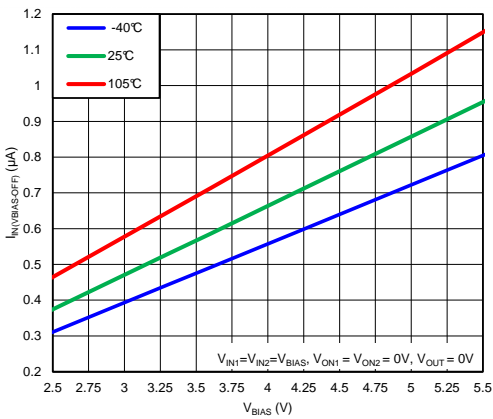


Figure 3. Shutdown Current vs. V_{BIAS} (Both Channels)

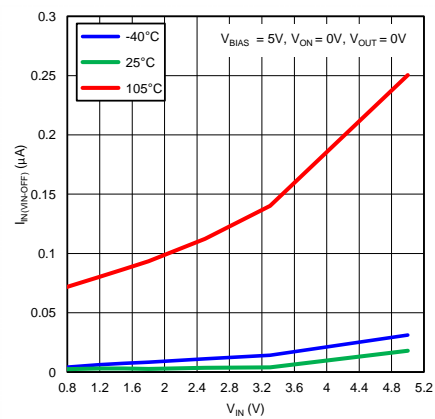


Figure 4. Off-State V_{IN} Current vs. V_{IN} (Single Channel)

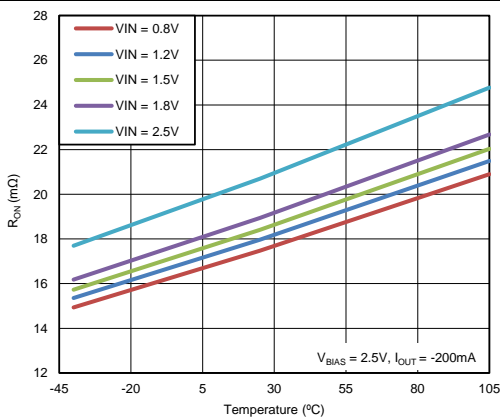


Figure 5. R_{ON} vs. Temperature ($V_{BIAS} = 2.5\text{ V}$, Single Channel)

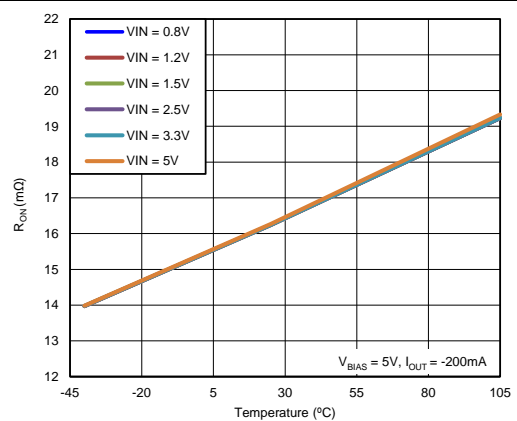


Figure 6. R_{ON} vs. Temperature ($V_{BIAS} = 5\text{ V}$, Single Channel)

Typical Characteristics (continued)

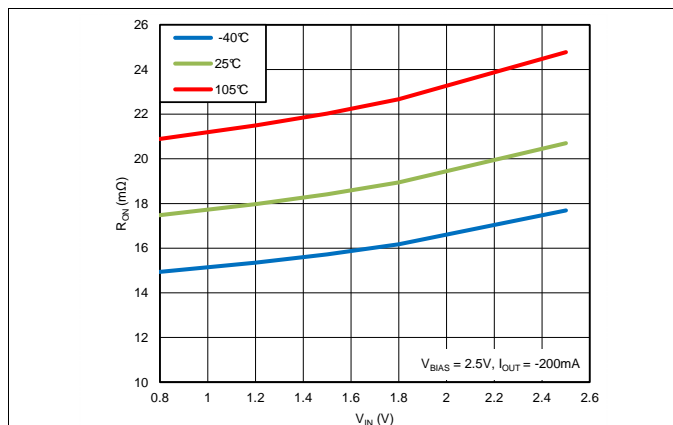


Figure 7. RON vs. VIN (VBIAS = 2.5 V, Single Channel)

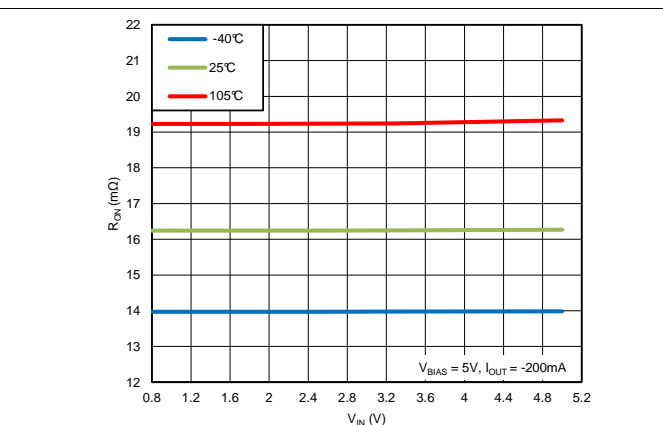


Figure 8. RON vs. VIN (VBIAS = 5 V, Single Channel)

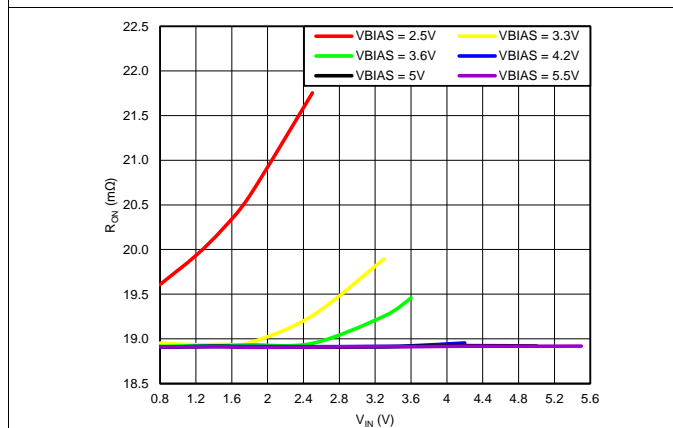


Figure 9. RON vs. VIN (TA = 25°C, Single Channel)

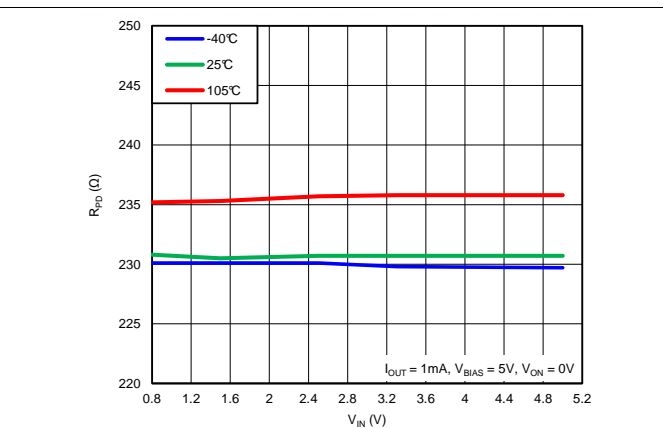


Figure 10. RPD vs. VIN (VBIAS = 5 V, Single Channel)

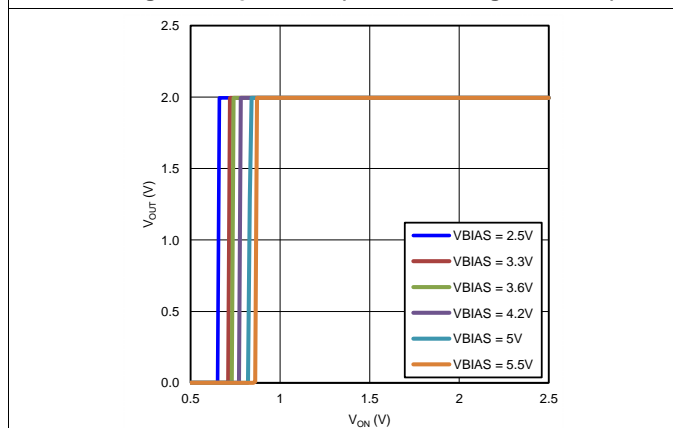


Figure 11. VOUT vs. VON (TA = 25°C, Single Channel)

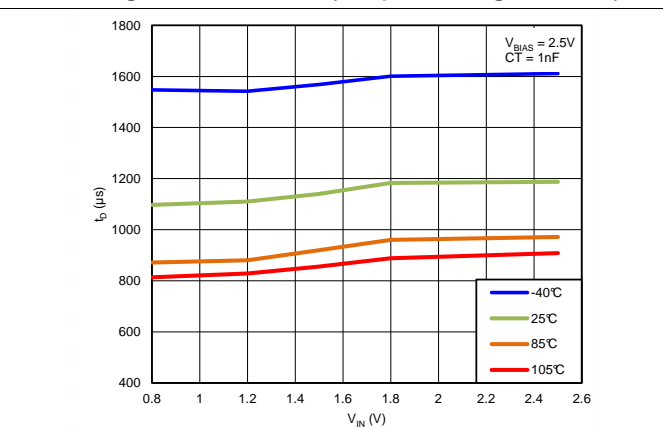


Figure 12. tD vs. VIN (VBIAS = 2.5 V, CT = 1 nF)

Typical Characteristics (continued)

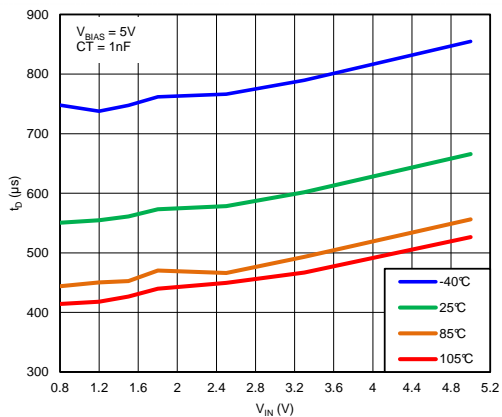


Figure 13. t_D vs. V_{IN} ($V_{BIAS} = 5\text{ V}$, $CT = 1\text{ nF}$)

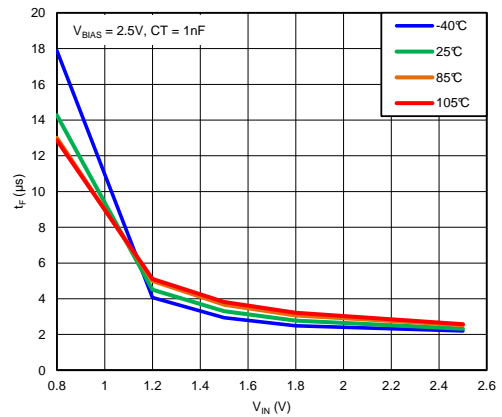


Figure 14. t_F vs. V_{IN} ($V_{BIAS} = 2.5\text{ V}$, $CT = 1\text{ nF}$)

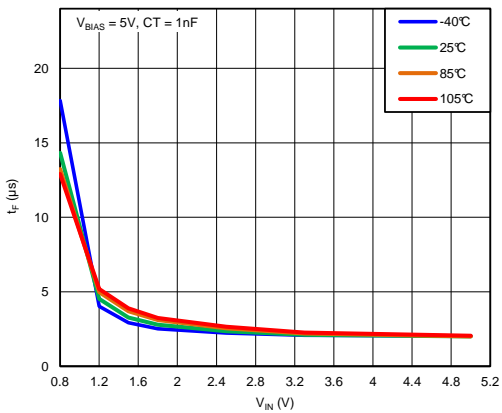


Figure 15. t_F vs. V_{IN} ($V_{BIAS} = 5\text{ V}$, $CT = 1\text{ nF}$)

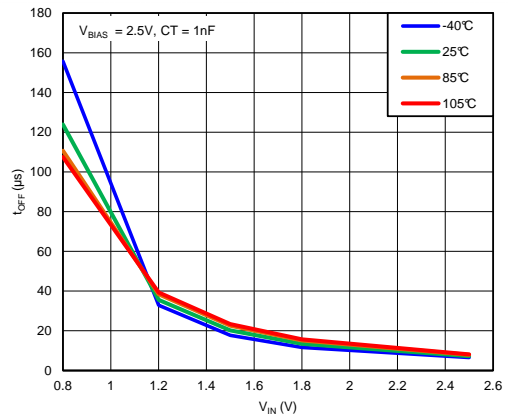


Figure 16. t_{OFF} vs. V_{IN} ($V_{BIAS} = 2.5\text{ V}$, $CT = 1\text{ nF}$)

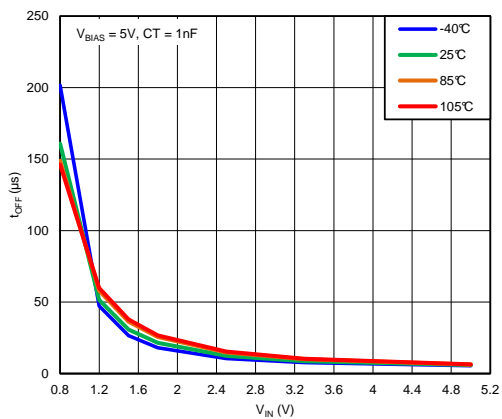


Figure 17. t_{OFF} vs. V_{IN} ($V_{BIAS} = 5\text{ V}$, $CT = 1\text{ nF}$)

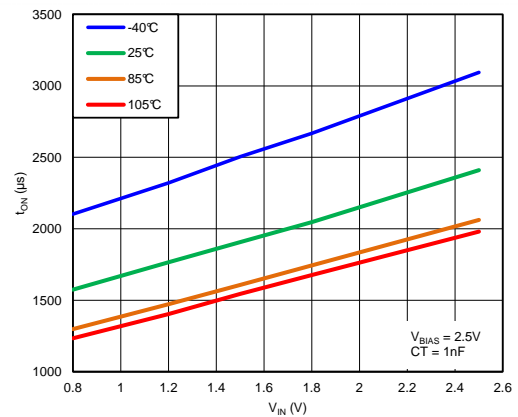


Figure 18. t_{ON} vs. V_{IN} ($V_{BIAS} = 2.5\text{ V}$, $CT = 1\text{ nF}$)

Typical Characteristics (continued)

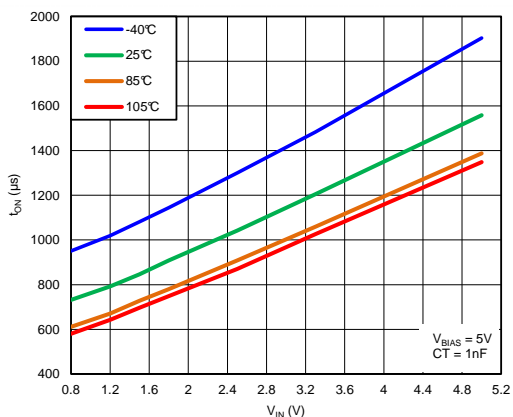


Figure 19. t_{ON} vs. V_{IN} ($V_{BIAS} = 5\text{ V}$, $CT = 1\text{ nF}$)

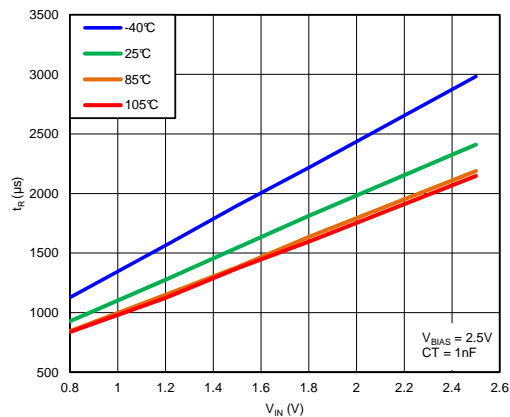


Figure 20. t_R vs. V_{IN} ($V_{BIAS} = 2.5\text{ V}$, $CT = 1\text{ nF}$)

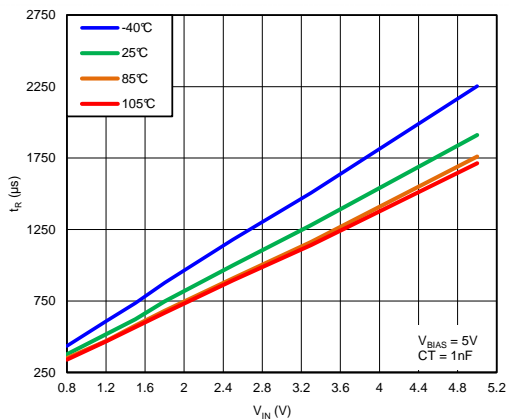


Figure 21. t_R vs. V_{IN} ($V_{BIAS} = 5\text{ V}$, $CT = 1\text{ nF}$)

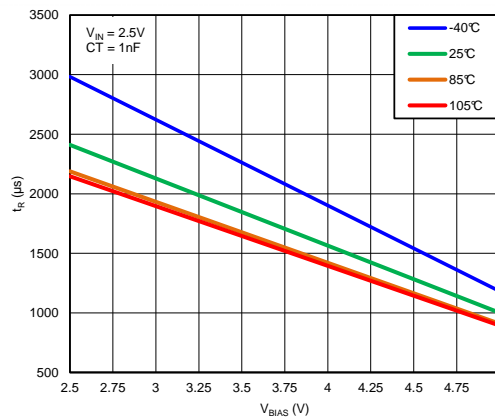


Figure 22. t_R vs. V_{BIAS} ($V_{IN} = 2.5\text{ V}$, $CT = 1\text{ nF}$)

6.9 Typical AC Characteristics

at $T_A = 25^\circ\text{C}$, $C_T = 1\text{ nF}$

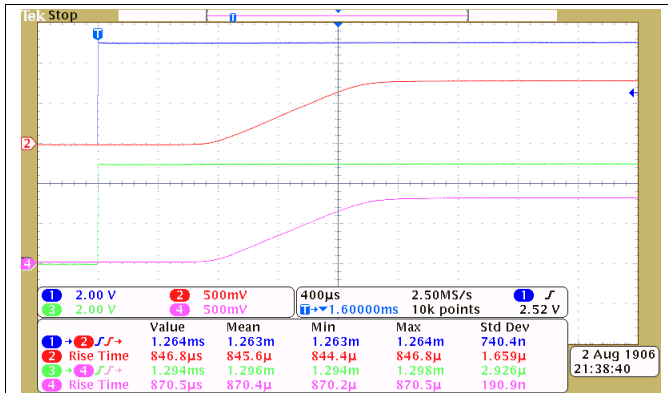


Figure 23. Turnon Response Time ($V_{IN} = 0.8\text{ V}$, $V_{BIAS} = 2.5\text{ V}$, $C_{IN} = 1\text{ }\mu\text{F}$, $C_L = 0.1\text{ }\mu\text{F}$, $R_L = 10\text{ }\Omega$)

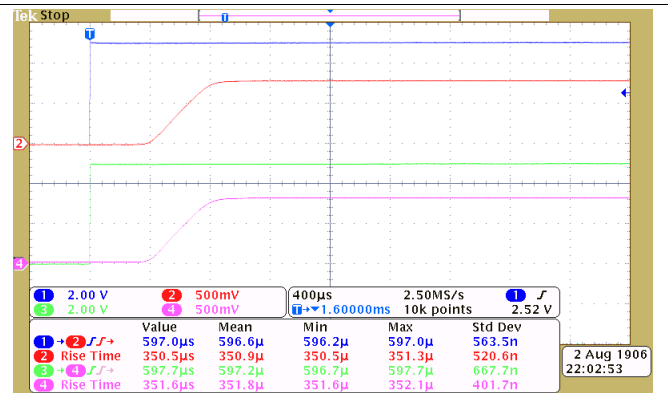


Figure 24. Turnon Response Time ($V_{IN} = 0.8\text{ V}$, $V_{BIAS} = 5\text{ V}$, $C_{IN} = 1\text{ }\mu\text{F}$, $C_L = 0.1\text{ }\mu\text{F}$, $R_L = 10\text{ }\Omega$)

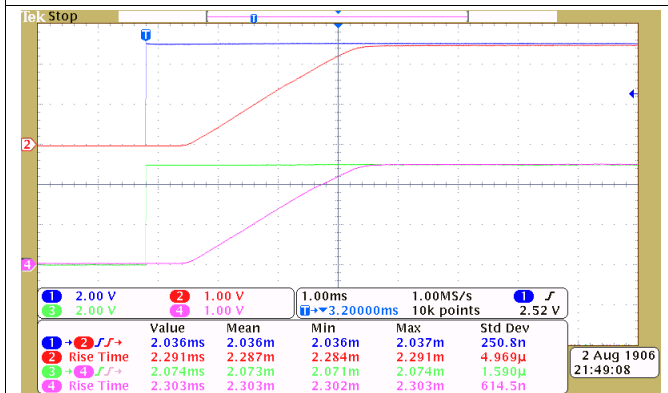


Figure 25. Turnon Response Time ($V_{IN} = 2.5\text{ V}$, $V_{BIAS} = 2.5\text{ V}$, $C_{IN} = 1\text{ }\mu\text{F}$, $C_L = 0.1\text{ }\mu\text{F}$, $R_L = 10\text{ }\Omega$)

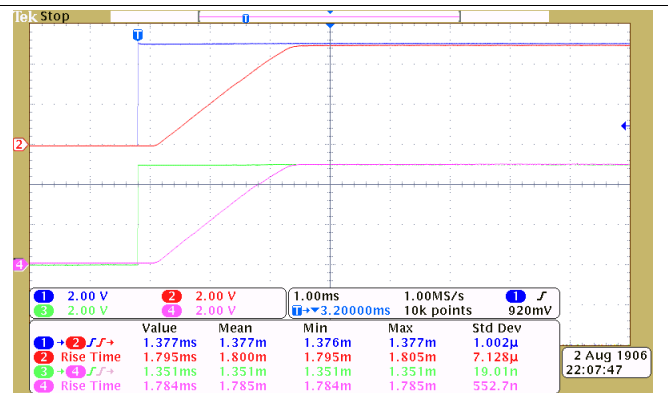


Figure 26. Turnon Response Time ($V_{IN} = 5\text{ V}$, $V_{BIAS} = 5\text{ V}$, $C_{IN} = 1\text{ }\mu\text{F}$, $C_L = 0.1\text{ }\mu\text{F}$, $R_L = 10\text{ }\Omega$)

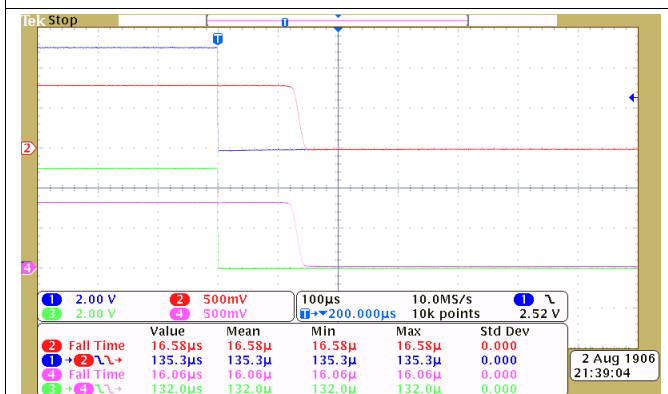


Figure 27. Turnoff Response Time ($V_{IN} = 0.8\text{ V}$, $V_{BIAS} = 2.5\text{ V}$, $C_{IN} = 1\text{ }\mu\text{F}$, $C_L = 0.1\text{ }\mu\text{F}$, $R_L = 10\text{ }\Omega$)

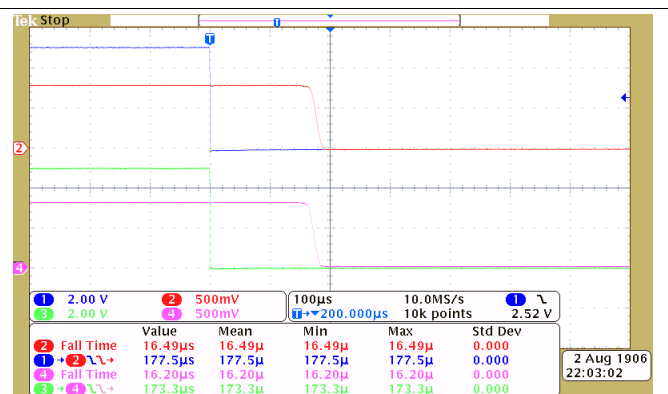
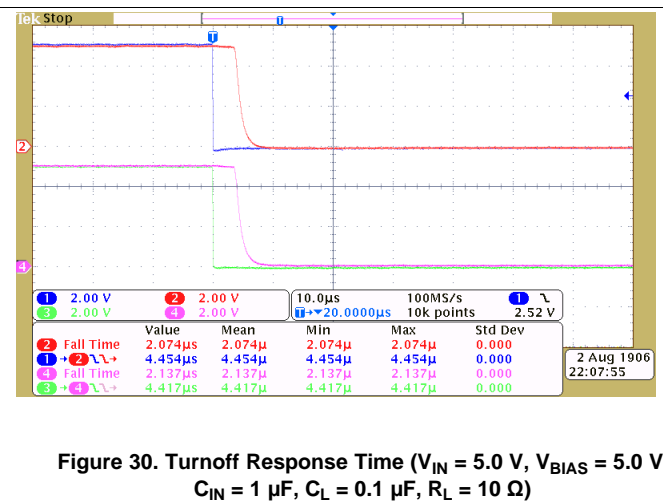
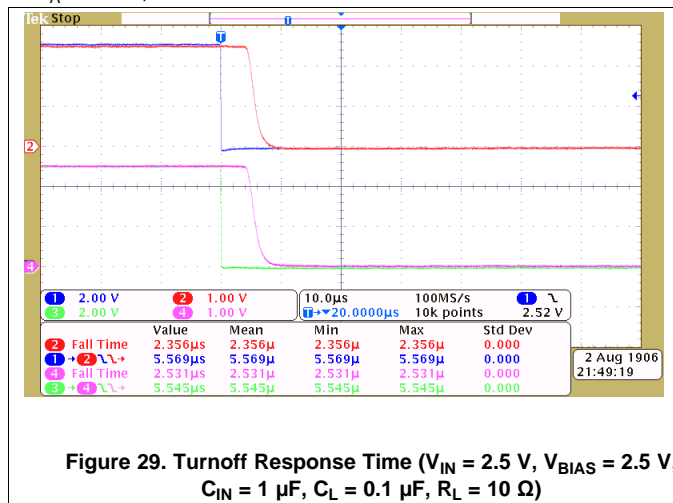


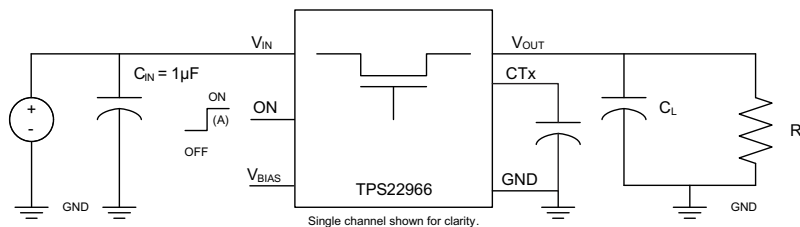
Figure 28. Turnoff Response Time ($V_{IN} = 0.8\text{ V}$, $V_{BIAS} = 5.0\text{ V}$, $C_{IN} = 1\text{ }\mu\text{F}$, $C_L = 0.1\text{ }\mu\text{F}$, $R_L = 10\text{ }\Omega$)

Typical AC Characteristics (continued)

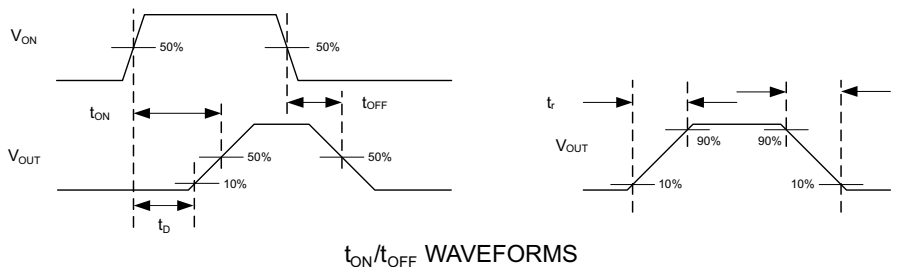
at $T_A = 25^\circ\text{C}$, $C_T = 1\text{ nF}$



7 Parameter Measurement Information



TEST CIRCUIT



(A) Rise and fall times of the control signal is 100ns.

Figure 31. Test Circuit and T_{ON}/T_{OFF} Waveforms

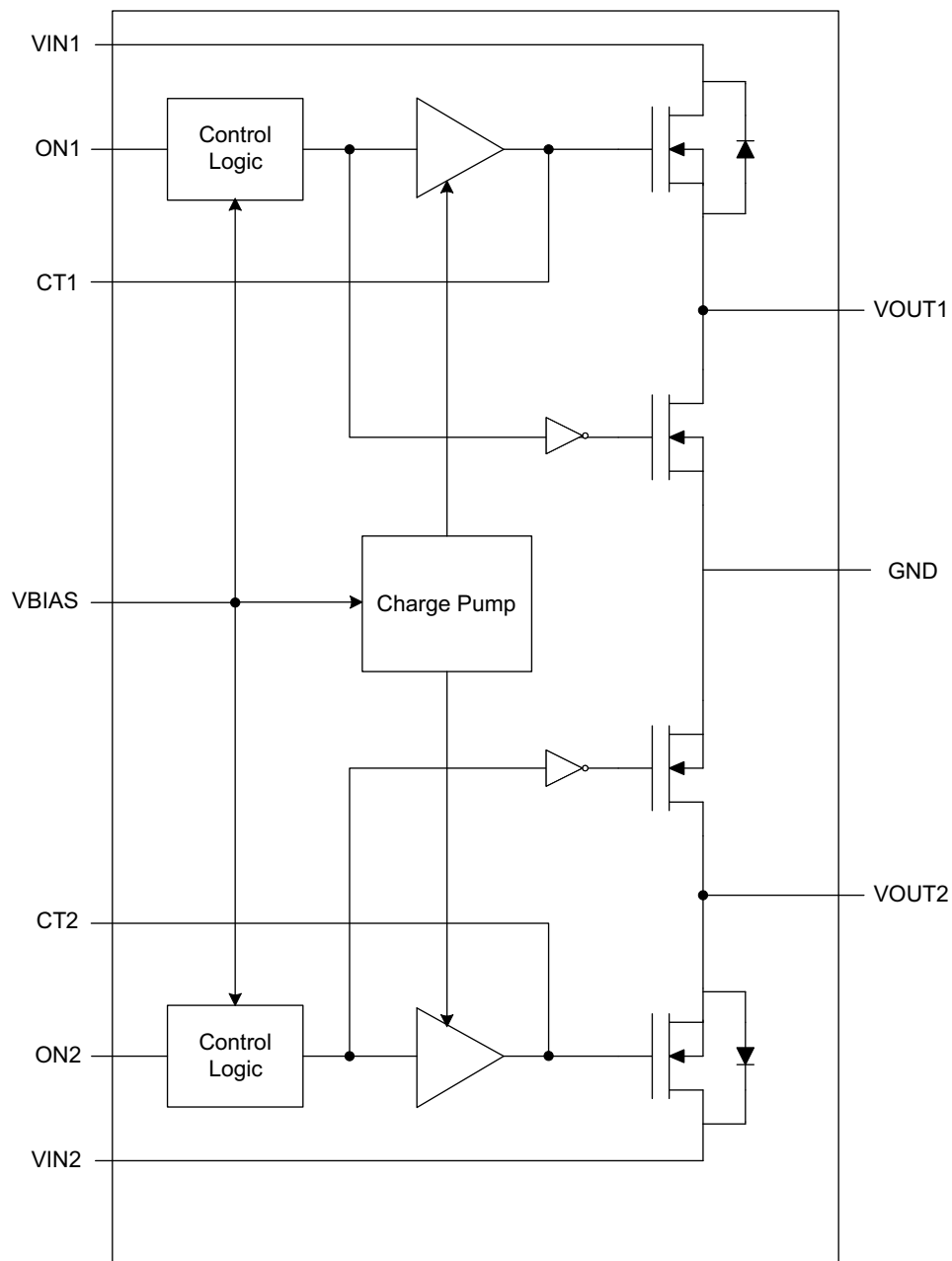
8 Detailed Description

8.1 Overview

The device is a dual-channel, 6-A load switch in a 14-terminal SON package. To reduce the voltage drop in high current rails, the device implements a low resistance N-channel MOSFET. The device has a programmable slew rate for applications that require specific rise-time.

The device has very low leakage current during off state. This prevents downstream circuits from pulling high standby current from the supply. Integrated control logic, driver, power supply, and output discharge FET eliminates the need for any external components, which reduces solution size and bill of materials (BOM) count.

8.2 Functional Block Diagram



8.3 Feature Description

8.3.1 ON/OFF Control

The ON pins control the state of the switch. Asserting ON high enables the switch. ON is active high and has a low threshold, making it capable of interfacing with low-voltage signals. The ON pin is compatible with standard GPIO logic threshold. It can be used with any microcontroller with 1.2-V or higher GPIO voltage. This pin cannot be left floating and must be tied either high or low for proper functionality.

8.3.2 Input Capacitor (Optional)

To limit the voltage drop on the input supply caused by transient inrush currents when the switch turns on into a discharged load capacitor, a capacitor needs to be placed between VIN and GND. A 1-μF ceramic capacitor, CIN, placed close to the pins, is usually sufficient. Higher values of CIN can be used to further reduce the voltage drop during high-current application. When switching heavy loads, it is recommended to have an input capacitor about 10 times higher than the output capacitor to avoid excessive voltage drop.

8.3.3 Output Capacitor (Optional)

Due to the integrated body diode in the NMOS switch, a CIN greater than CL is highly recommended. A CL greater than CIN can cause VOUT to exceed VIN when the system supply is removed. This could result in current flow through the body diode from VOUT to VIN. A CIN to CL ratio of 10 to 1 is recommended for minimizing VIN dip caused by inrush currents during startup, however a 10 to 1 ratio for capacitance is not required for proper functionality of the device. A ratio smaller than 10 to 1 (such as 1 to 1) could cause slightly more VIN dip upon turn-on due to inrush currents. This can be mitigated by increasing the capacitance on the CT pin for a longer rise time (see Figure 4).

8.3.4 VIN and VBIAS Voltage Range

For optimal RON performance, make sure VIN ≤ VBIAS. The device will still be functional if VIN > VBIAS but it will exhibit RON greater than what is listed in *Electrical Characteristics*. See Figure 32 for an example of a typical device. Notice the increasing RON as VIN exceeds VBIAS voltage. Be sure to never exceed the maximum voltage rating for VIN and VBIAS.

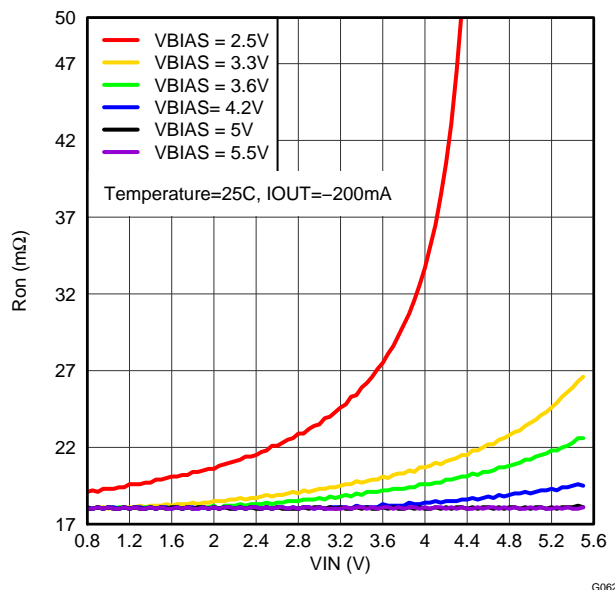


Figure 32. RON vs VIN (VIN > VBIAS, Single Channel)

8.4 Device Functional Modes

Table 1. Functions Table

ONx	VINx to VOUTx	VOUTx to GND
L	Off	On
H	On	Off

9 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

This application demonstrates how the TPS22966 can be used to limit inrush current when powering on downstream modules.

9.2 Typical Application

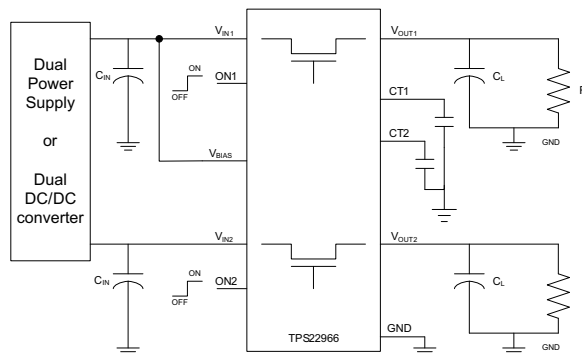


Figure 33. Typical Application Circuit

9.2.1 Design Requirements

Table 2. Design Parameters

DESIGN PARAMETER	VALUE
Input voltage	3.3 V
Bias voltage	5 V
Load capacitance (C _L)	22 μF
Maximum acceptable inrush current	400 mA

9.2.2 Detailed Design Procedure

When the switch is enabled, the output capacitors must be charged up from 0 V to the set value (3.3 V in this example). This charge arrives in the form of inrush current. Inrush current can be calculated using [Equation 1](#):

$$\text{Inrush Current} = C \times dV/dt$$

where

- C = output capacitance
- dV = output voltage
- dt = rise time

(1)

The TPS22966 offers adjustable rise time for VOUT. This feature allows the user to control the inrush current during turnon. The appropriate rise time can be calculated using [Table 2](#) and the inrush current equation.

$$400 \text{ mA} = 22 \text{ μF} \times 3.3 \text{ V/dt}$$

(2)

$$dt = 181.5 \text{ μs}$$

(3)

To ensure an inrush current of less than 400 mA, choose a CT value that will yield a rise time of more than 181.5 μs. See the oscilloscope captures in [Application Curves](#) for an example of how the CT capacitor can be used to reduce inrush current.

9.2.2.1 Adjustable Rise Time

A capacitor to GND on the CTx pins sets the slew rate for each channel. To ensure desired performance, a capacitor with a minimum voltage rating of 25 V should be used on the CTx pin. An approximate formula for the relationship between CTx and slew rate is (the equation below accounts for 10% to 90% measurement on V_{OUT} and does **NOT** apply for CTx = 0 pF. Use [Table 3](#) to determine rise times for when CTx = 0 pF):

$$SR = 0.32 \times CT + 13.7$$

where

- SR = slew rate (in $\mu\text{s/V}$)
- CT = the capacitance value on the CTx pin (in pF)
- The units for the constant 13.7 is in $\mu\text{s/V}$. (4)

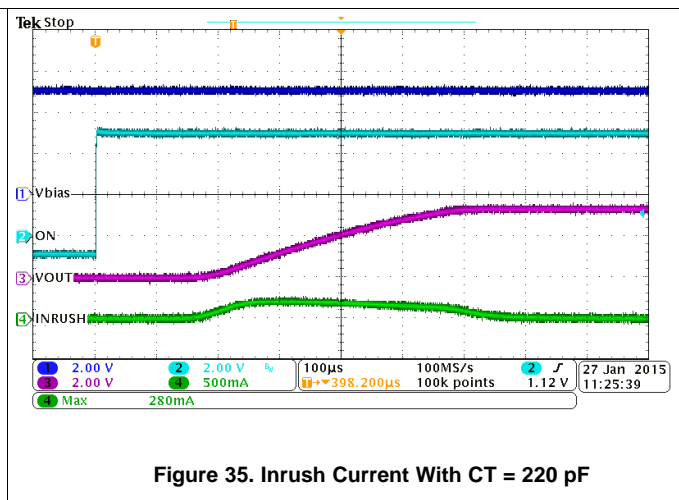
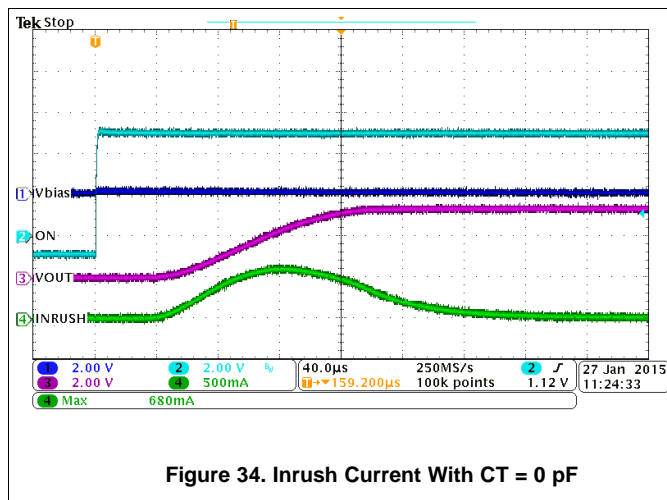
Rise time can be calculated by multiplying the input voltage by the slew rate. [Table 3](#) shows rise time values measured on a typical device. Rise times shown below are only valid for the power-up sequence where V_{IN} and V_{BIAS} are already in steady state condition, and the ON pin is asserted high.

Table 3. Rise Time Values

CTx (pF)	RISE TIME (μs) 10% - 90%, C _L = 0.1 μF , C _{IN} = 1 μF , R _L = 10 Ω TYPICAL VALUES at 25°C, V _{BIAS} = 5V, 25V X7R 10% CERAMIC CAP						
	5V	3.3V	1.8V	1.5V	1.2V	1.05V	0.8V
0	124	88	63	60	53	49	42
220	481	323	193	166	143	133	109
470	855	603	348	299	251	228	175
1000	1724	1185	670	570	469	411	342
2200	3328	2240	1308	1088	893	808	650
4700	7459	4950	2820	2429	1920	1748	1411
10000	16059	10835	6040	5055	4230	3770	3033

9.2.3 Application Curves

V_{BIAS} = 5 V ; V_{IN} = 3.3 V ; C_L = 22 μF



10 Power Supply Recommendations

The device is designed to operate from a VBIAS range of 2.5 V to 5.5 V and a VIN range of 0.8 V to VBIAS.

11 Layout

11.1 Layout Guidelines

For best performance, all traces should be as short as possible. To be most effective, the input and output capacitors should be placed close to the device to minimize the effects that parasitic trace inductances may have on normal operation. Using wide traces for VIN, VOUT, and GND helps minimize the parasitic electrical effects along with minimizing the case to ambient thermal impedance.

11.2 Layout Example

Notice the thermal vias located under the exposed thermal pad of the device. This allows for thermal diffusion away from the device.

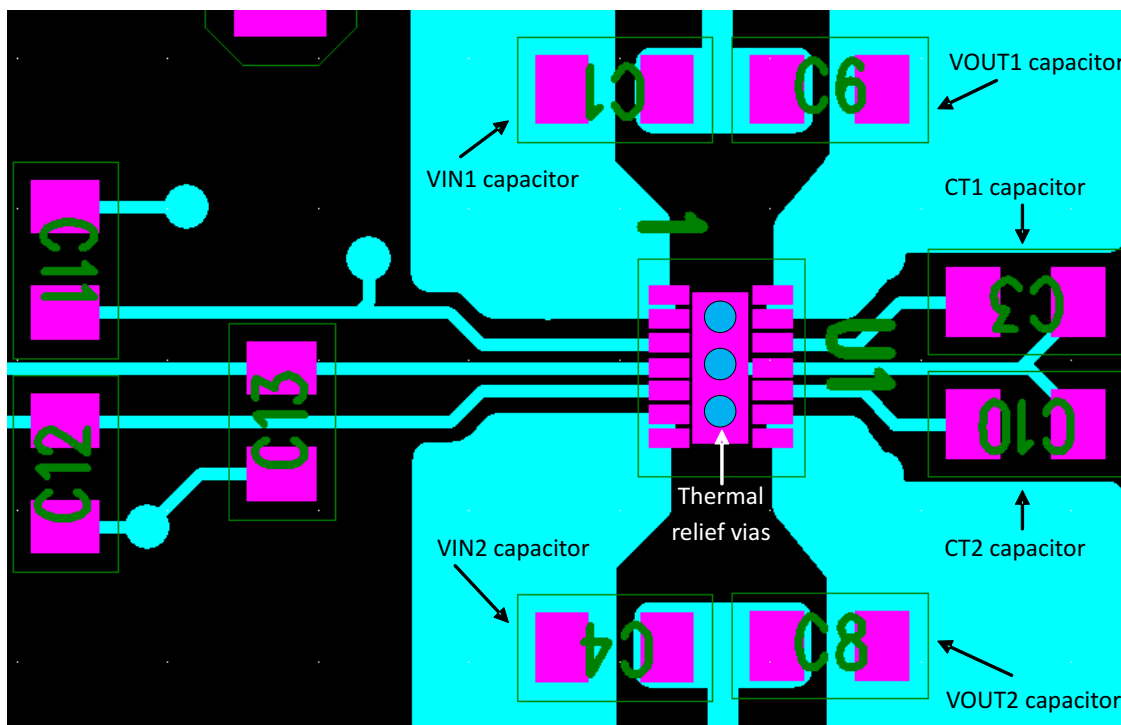


Figure 36. PCB Layout Example

11.3 Power Dissipation

The maximum IC junction temperature should be restricted to 125°C under normal operating conditions. To calculate the maximum allowable power dissipation, $P_{D(max)}$ for a given output current and ambient temperature, use the following equation:

$$P_{D(max)} = \frac{T_{J(max)} - T_A}{\theta_{JA}}$$

where

- $P_{D(max)}$ = maximum allowable power dissipation
- $T_{J(max)}$ = maximum allowable junction temperature (125°C for the TPS22966)
- T_A = ambient temperature of the device
- θ_{JA} = junction to air thermal impedance. See [Thermal Information](#). This parameter is highly dependent upon board layout. (5)

12 Device and Documentation Support

12.1 Trademarks

Ultrabook is a trademark of Intel.

All other trademarks are the property of their respective owners.

12.2 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

12.3 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TPS22966DPUR	ACTIVE	WSO8	DPU	14	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 105	RB966	Samples
TPS22966DPUT	ACTIVE	WSO8	DPU	14	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 105	RB966	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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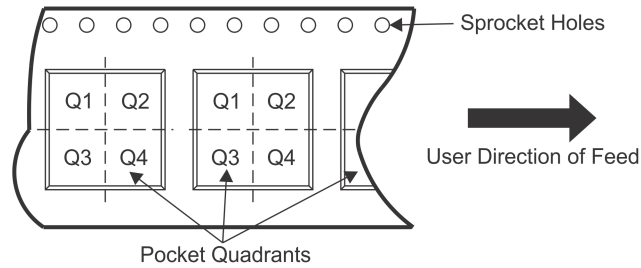
OTHER QUALIFIED VERSIONS OF TPS22966 :

- Automotive: [TPS22966-Q1](#)

NOTE: Qualified Version Definitions:

- Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS22966DPUR	WSON	DPU	14	3000	180.0	8.4	2.25	3.25	1.05	4.0	8.0	Q1
TPS22966DPUT	WSON	DPU	14	250	180.0	8.4	2.25	3.25	1.05	4.0	8.0	Q1

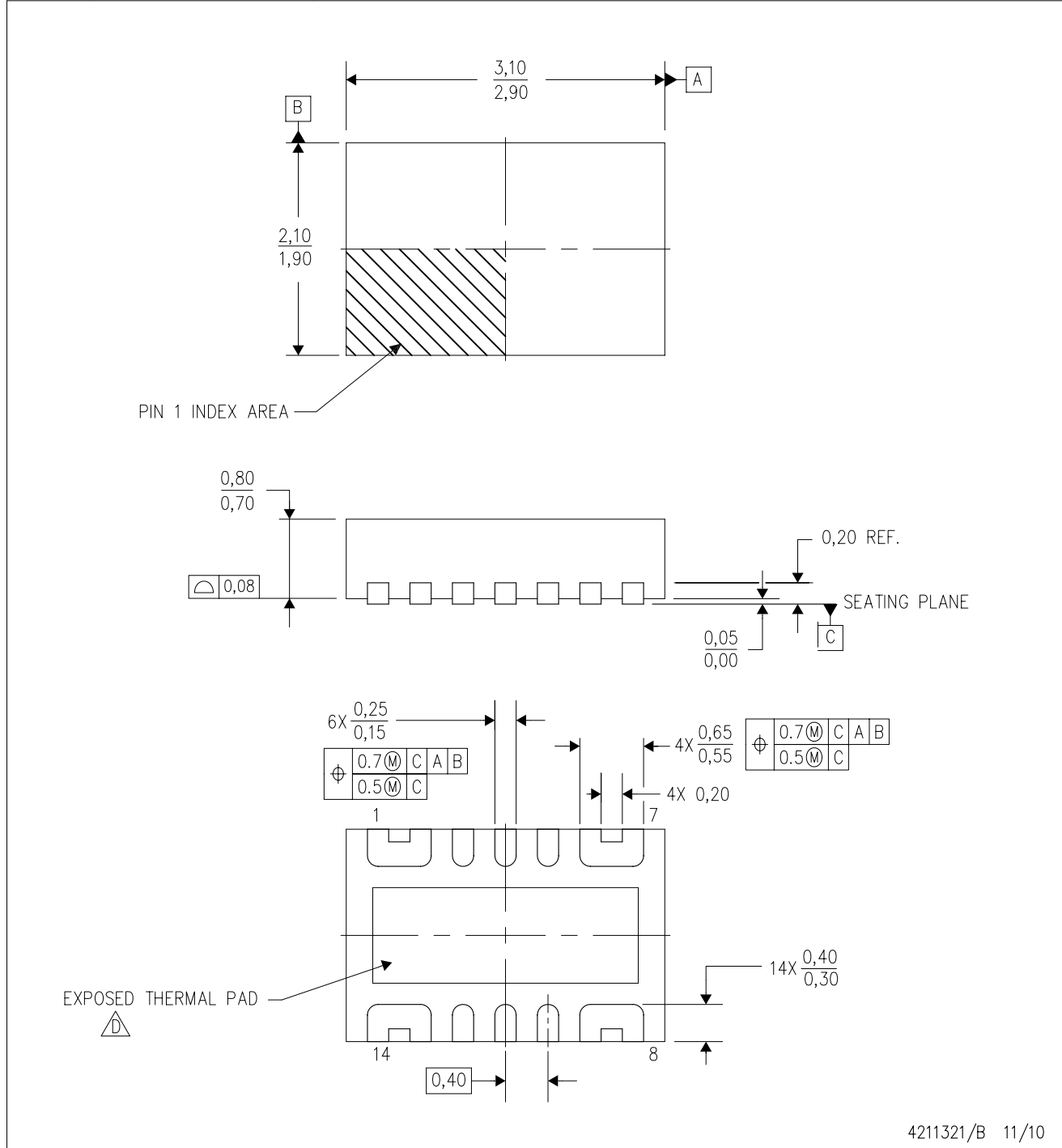
TAPE AND REEL BOX DIMENSIONS


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS22966DPUR	WSON	DPU	14	3000	210.0	185.0	35.0
TPS22966DPUT	WSON	DPU	14	250	210.0	185.0	35.0

DPU (R-PWSON-N14)

PLASTIC SMALL OUTLINE NO-LEAD



- NOTES:
- All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
 - This drawing is subject to change without notice.
 - Small Outline No-Lead (SON) package configuration.
 - The package thermal pad must be soldered to the board for thermal and mechanical performance. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.
 - This package is Pb-free.

THERMAL PAD MECHANICAL DATA

DPU (R-PWSON-N14)

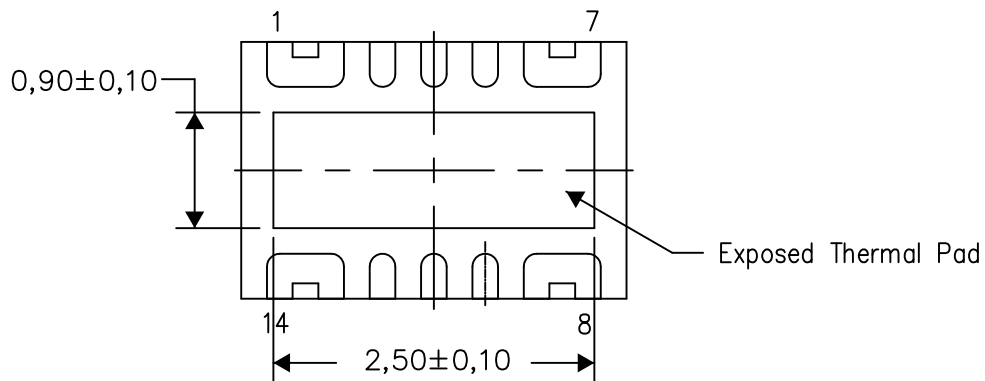
PLASTIC SMALL OUTLINE NO-LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Bottom View

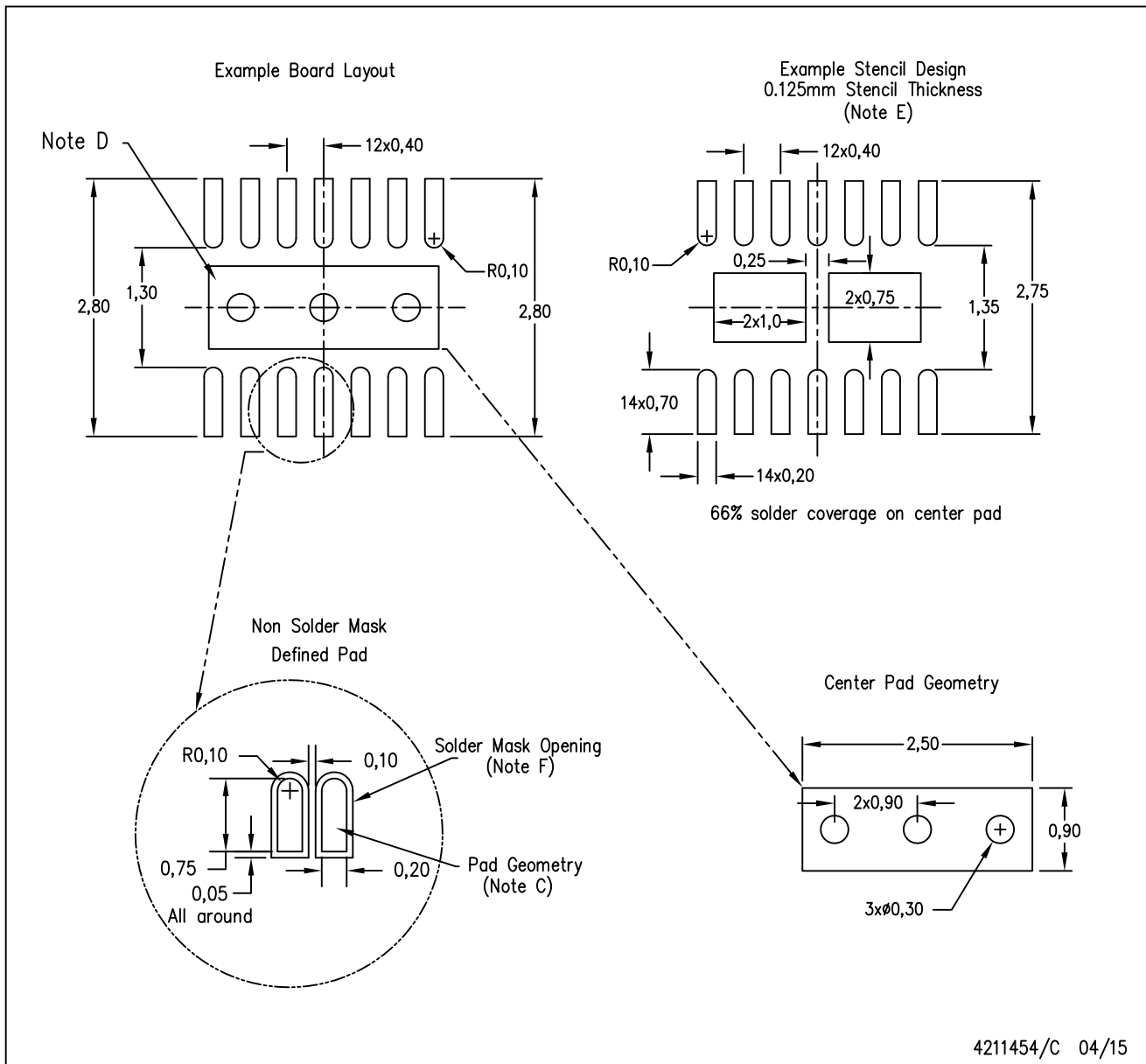
Exposed Thermal Pad Dimensions

4211395/C 04/15

NOTE: All linear dimensions are in millimeters

DPU (R-PWSON-N14)

PLASTIC SMALL OUTLINE NO-LEAD



- NOTES:
- All linear dimensions are in millimeters.
 - This drawing is subject to change without notice.
 - Publication IPC-7351 is recommended for alternate designs.
 - This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <<http://www.ti.com>>.
 - Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
 - Customers should contact their board fabrication site for minimum solder mask web tolerances between signal pads.

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